Transmission eigenvalue problems with sign-changing coefficients

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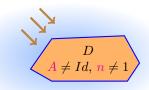
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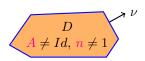




Scattering in time-harmonic regime by an inclusion D (coefficients A and n) in \mathbb{R}^2 : we look for an incident wave that does not scatter.

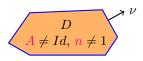


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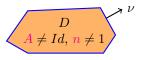
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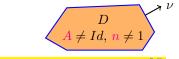
$$\begin{array}{rcl} \operatorname{div}\left(\mathbf{A}\nabla u\right) + k^2\mathbf{n}u & = & 0 & \text{ in } D \\ \Delta w + k^2w & = & 0 & \text{ in } D \end{array}$$



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$$\begin{array}{ccccc} \operatorname{div}\left(A\nabla u\right) + k^2 n u & = & 0 & \operatorname{in} D \\ \Delta w + k^2 w & = & 0 & \operatorname{in} D \\ u - w & = & 0 & \operatorname{on} \partial D \\ \nu \cdot A\nabla u - \nu \cdot \nabla w & = & 0 & \operatorname{on} \partial D. \end{array}$$



Transmission conditions on ∂D

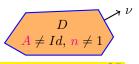
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Find
$$(u, w) \in H^1(D) \times H^1(D)$$
 such that: $\operatorname{div}(A\nabla u) + k^2 n u = 0$ in D

$$\frac{\Delta w + k^2 w}{u + w} = 0$$
 in D

$$\frac{u - w}{v \cdot A\nabla u - v \cdot \nabla w} = 0$$
 on ∂D .



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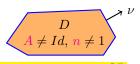
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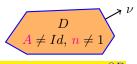
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DEFINITION. Values of $k \in \mathbb{C}$ for which this problem has a nontrivial solution (u, w) are called transmission eigenvalues.

 \blacktriangleright The goal in this talk is to prove that the set of transmission eigenvalues is at most discrete.

▶ k is a transmission eigenvalue if and only if there exists $(u, w) \in X \setminus \{0\}$ such that, for all $(u', w') \in X$,

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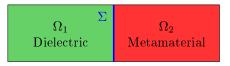
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Idea 1: Analogy with another non standard transmission problem ...

Time-harmonic problem in electromagnetism (at a given frequency) set in a heterogeneous bounded domain Ω of \mathbb{R}^2 :



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 \triangleright Eigenvalue problem for E_z in 2D:

Find
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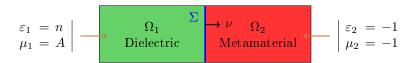
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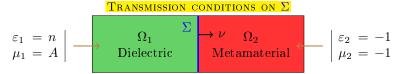
$$\int_{\Omega_1} \mu_1^{-1} \nabla v \cdot \overline{\nabla v'} - \int_{\Omega_2} |\mu_2|^{-1} \nabla v \cdot \overline{\nabla v'} = k^2 \left(\int_{\Omega_1} \varepsilon_1 v \overline{v'} - \int_{\Omega_2} |\varepsilon_2| v \overline{v'} \right).$$

$$_{4/15}$$

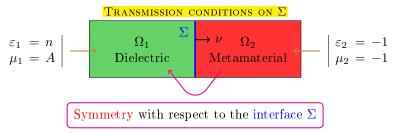
DMTEP in the domain Ω :



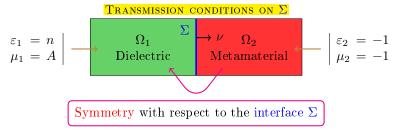
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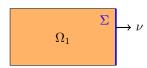
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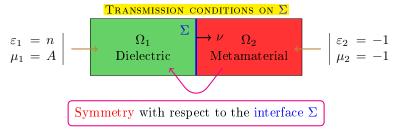
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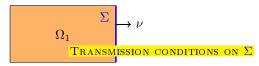
• We obtain a problem analogous to the ITEP in Ω_1 :



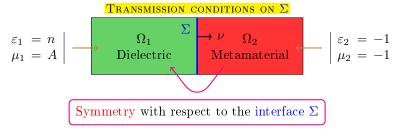
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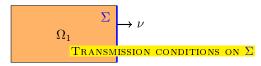
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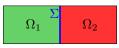
▶ The interface Σ in the DMTEP plays the role of the boundary ∂D in the ITEP.

Outline of the talk: three steps

- 1 An analogy between two transmission problems
- 2 The T-coercivity method for the Dielectric/Metamaterial Transmission Problem
- 3 The T-coercivity method for the Interior Transmission Problem

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$$\mu_1 > 0$$
 | Ω_1 Ω_2 | $\mu_2 < 0$ Ω_2 Ω_2 | Ω_2 |

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Idea 2: Use the T-coercivity approach to deal with problem (\mathscr{P}_V) .

$$(\mathscr{P}_V) \mid \operatorname{Find} v \in \mathrm{H}^1_0(\Omega) \text{ such that:} \\ a(v,v') = l(v'), \, \forall v' \in \mathrm{H}^1_0(\Omega).$$

$$(\mathscr{P}_V) \Leftrightarrow (\mathscr{P}_V^{\mathsf{T}}) \ \middle| \ \text{Find} \ v \in \mathrm{H}^1_0(\Omega) \ \text{such that:} \\ a(v, \mathsf{T} v') = l(\mathsf{T} v'), \ \forall v' \in \mathrm{H}^1_0(\Omega).$$

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Goal: Find T such that
$$a$$
 is T-coercive:
$$\int_{\Omega} \mu^{-1} \nabla v \cdot \nabla(\mathsf{T} v) \geq C \|v\|_{\mathrm{H}_{0}^{1}(\Omega)}^{2}.$$
 In this case, Lax-Milgram $\Rightarrow (\mathscr{P}_{V}^{\mathsf{T}})$ (and so (\mathscr{P}_{V})) is well-posed.

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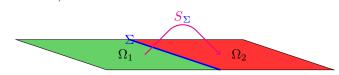
$$\begin{array}{|c|c|c|c|c|} \hline \textbf{1} & \text{Define } \mathbf{T}_1 \, v = \begin{vmatrix} v_1 & & \text{in } \Omega_1 \\ -v_2 + \dots & & \text{in } \Omega_2 \end{vmatrix}$$

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1 Define $T_1 v = \begin{vmatrix} v_1 & \text{in } \Omega_1 \\ -v_2 + 2S_{\Sigma}v_1 & \text{in } \Omega_2 \end{vmatrix}$, where S_{Σ} is the symmetry.

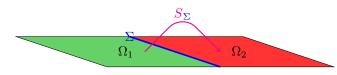


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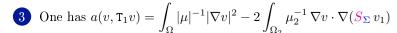
$$(\mathscr{P}_V) \Leftrightarrow (\mathscr{P}_V^{\mathsf{T}}) \mid \text{Find } v \in \mathrm{H}_0^1(\Omega) \text{ such that:} \\ a(v, \mathsf{T}v') = l(\mathsf{T}v'), \, \forall v' \in \mathrm{H}_0^1(\Omega).$$

Goal: Find T such that a is T-coercive: $\int_{\Omega} \mu^{-1} \nabla v \cdot \nabla(\mathsf{T} v) \geq C \|v\|_{\mathrm{H}_{0}^{1}(\Omega)}^{2}.$ In this case, Lax-Milgram $\Rightarrow (\mathscr{P}_{V}^{\mathsf{T}})$ (and so (\mathscr{P}_{V})) is well-posed.

1 Define $T_1 v = \begin{vmatrix} v_1 & \text{in } \Omega_1 \\ -v_2 + 2S_{\Sigma}v_1 & \text{in } \Omega_2 \end{vmatrix}$, where S_{Σ} is the symmetry.



2 $T_1 \circ T_1 = Id$ so T_1 is an isomorphism of $H_0^1(\Omega)$



3 One has $a(v, \mathsf{T}_1 v) = \int_{\Omega} |\mu|^{-1} |\nabla v|^2 - 2 \int_{\Omega_2} \mu_2^{-1} |\nabla v| \cdot \nabla (S_{\Sigma} v_1)$

Young's inequality $+ \|S_{\Sigma}\| = 1 \Rightarrow a$ is T-coercive when $|\mu_2| > \mu_1$.

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- Young's inequality $+ \|S_{\Sigma}\| = 1 \Rightarrow a$ is T-coercive when $|\mu_2| > \mu_1$.
- 4 With $\mathtt{T}_2 v = \left| \begin{array}{cc} v_1 2S_{\Sigma}v_2 & \text{in } \Omega_1 \\ -v_2 & \text{in } \Omega_2 \end{array} \right., \quad a \text{ is } \text{\mathbf{T}-coercive when } \mu_1 > |\mu_2| \ .$

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- **5** Conclusion:

THEOREM. The operator div $(\mu^{-1} \nabla \cdot)$ is an isomorphism from $H_0^1(\Omega)$ to $H^{-1}(\Omega)$ if and only if the contrast $\kappa_{\mu} = \mu_1/\mu_2$ satisfies $\kappa_{\mu} \neq -1$.

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▶ By a localization process, when μ_1 and μ_2 are not constant, we can prove that div $(\mu^{-1} \nabla \cdot)$ is of Fredholm type when

$$\inf_{\Omega_1 \cap \mathcal{V}} \mu_1 / \inf_{\Omega_2 \cap \mathcal{V}} \mu_2 < -1 \qquad \text{or} \qquad \sup_{\Omega_1 \cap \mathcal{V}} \mu_1 / \sup_{\Omega_2 \cap \mathcal{V}} \mu_2 > -1$$

where \mathcal{V} is a neighbourhood of Σ .

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This technique also allows to deal with non symmetric configurations.

1 An analogy between two transmission problems

2 The T-coercivity method for the Dielectric/Metamaterial Transmission Problem

3 The T-coercivity method for the Interior Transmission Problem

 \triangleright Define on X × X the sesquilinear form

$$a((u,w),(u',w')) = \int_{\Omega} A \nabla u \cdot \overline{\nabla u'} - \nabla w \cdot \overline{\nabla w'} - k^2 (nu\overline{u'} - w\overline{w'}),$$

with
$$X = \{(u, w) \in H^1(\Omega) \times H^1(\Omega) | u - w \in H^1_0(\Omega) \}.$$

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Introduce the isomorphism T(u, w) = (u - 2w, -w).

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with $X = \{(u, w) \in H^1(\Omega) \times H^1(\Omega) \mid u - w \in H^1_0(\Omega)\}.$

- Introduce the isomorphism T(u, w) = (u 2w, -w).
- For $k \in \mathbb{R}i \setminus \{0\}$, A > Id and n > 1, one finds

$$\Re e \, a((u,w), \mathsf{T}(u,w)) \ge C \, (\|u\|_{\mathrm{H}^1(\Omega)}^2 + \|w\|_{\mathrm{H}^1(\Omega)}^2), \quad \forall (u,w) \in \mathsf{X}.$$

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PROPOSITION. Suppose that A > Id and n > 1. Then the set of transmission eigenvalues is discrete and countable.

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PROPOSITION. Suppose that A > Id and n > 1. Then the set of transmission eigenvalues is discrete and countable.

This result can be extended to situations where A - Id and n - 1 change sign in Ω working with $T(u, w) = (u - 2\chi w, w)$.

When A = Id, the ITP is not of Fredholm type in X likewise the DMTP is not of Fredholm type in $H_0^1(\Omega)$ when $\mu_1 = -\mu_2$.

We change the functional framework working on the difference $v := u - w \in H_0^2(D)$: k is a transmission eigenvalue if and only if there exists $v \in H_0^2(D) \setminus \{0\}$ such that, for all $v' \in H_0^2(D)$,

$$\int_{D} \frac{1}{1-n} (\Delta v + k^{2} n v) (\Delta v' + k^{2} v') = 0.$$

► We focus on the principal part:

$$(\mathscr{F}_V) \left| \begin{array}{l} \text{Find } v \in \mathrm{H}^2_0(D) \text{ such that:} \\ \underbrace{\int_D \frac{1}{1-n} \Delta v \Delta v'}_{a(v,v')} = \underbrace{\langle f, v' \rangle_D}_{l(v')}, \quad \forall v' \in \mathrm{H}^2_0(D). \end{array} \right|$$

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$$\underbrace{\begin{array}{c} F_{V} \\ \left(\mathscr{F}_{V}\right) \end{array}} \left[\begin{array}{c} F_{D} \frac{1}{1-n} \Delta v \Delta v' = \underbrace{\left\langle f, v' \right\rangle_{D}}, \quad \forall v' \in \mathcal{H}_{0}^{2}(D). \end{array}\right]}_{a(v,v')} V' \in \mathcal{H}_{0}^{2}(D).$$

We change the functional framework working on the difference $v := u - w \in H_0^2(D)$: k is a transmission eigenvalue if and only if there exists $v \in \mathrm{H}_0^2(D) \setminus \{0\}$ such that, for all $v' \in \mathrm{H}_0^2(D)$,

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 $\int_{D} \frac{1}{1-n} (\Delta v + k^2 n v) (\Delta v' + k^2 v') = 0.$ We focus on the principal part: $\int_{POSSIBLE} \frac{T_{RANSMISSION\ PROBLEM\ WITH\ A}}{POSSIBLE\ SIGN\ CHANGING\ COEFFICIENT}$ $(\mathscr{F}_{V}) \left| \underbrace{\int_{D} \frac{1}{1-n} \Delta v \Delta v'}_{a(v,v')} = \underbrace{\langle f,v' \rangle_{D}}_{l(v')}, \quad \forall v' \in \mathrm{H}^{2}_{0}(D). \right|$



Idea 3: This transmission problem is very different from DMTP.

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Idea 3: This transmission problem is very different from DMTP.

THEOREM. The problem (\mathscr{F}_V) is well-posed in the Fredholm sense as soon as 1-n does not change sign in a neighbourhood of ∂D .

Proof: T-coercivity or see J. Sylvester's work for a more precise study. $_{13/15}$

Generalizations

- T-coercivity approach can be used for non-constant coefficients (L^{∞}) and other problems (Maxwell's equations, elasticity, ...).
- ✓ It allows to justify the convergence of standard finite element methods.
- \spadesuit What happens when A Id change sign in a neighbourhood of the boundary?
 - ► For the equivalent DMTP, strong singularities appear at the interface and H¹ is no longer the appropriate functional framework. We observe a black hole phenomenon (joint work with X. Claeys).
- ♠ We are not able to use the T-coercivity technique to prove existence of transmission eigenvalues.
 - \Rightarrow T-coercivity gives positivity but operators are no longer symmetric.

Thank you for your attention.

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